

System Identification and Control Using Quadratic Neural Networks

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Abstract—This paper proposes convex formulations of system identification and control for nonlinear systems using two layer quadratic neural networks. The results in the paper cast system identification, stability and control design as convex optimization problems, which can be solved efficiently with polynomial-time algorithms. The main advantage of using quadratic neural networks for system identification and control as opposed to other neural networks is the fact that they provide a smooth (quadratic) mapping between the input and the output of the network. This allows one to cast stability and control for quadratic neural network models as a Sum of Squares (SOS) optimization, which is a convex optimization program that can be efficiently solved. Additionally, these networks offer other advantages, such as the fact that the architecture is a by-product of the design and is not determined a-priori, and the training can be done by solving a convex optimization problem so that the global optimum of the weights is achieved. It also appears from the examples in this paper that quadratic networks work extremely well using only a small fraction of the training data.

I. INTRODUCTION

Artificial neural networks have been an active area of research with applications such as image recognition, natural language processing, and signal processing to name a few. They have also been applied to system identification as a black-box model [1]. However, system identification using neural networks does not typically lead to a smooth analytical model. Additionally, neural networks did not yet reach a wide use in safety-critical applications of control systems, such as autonomous vehicles, because of the lack of theoretical guarantees on safety, stability, and performance, which are of prime concern for such applications. In particular, formal results on Lyapunov stability of a system in feedback with a neural network controller are very scarce and only started appearing very recently in the literature [2], [3]. One main difficulty to obtain provable stability guarantees when using neural networks is the fact that the mapping between the input and the output of the network does not have a concise analytical expression and/or is not smooth. Additionally, for most neural networks one must decide on the network architecture before the training of the network can be performed. This is most often a difficult trial-and-error task that is heavily dependent on the application, although upper bounds on the number of neurons in feedforward networks needed to learn a given number of data were determined in reference [4]. Furthermore, once an architecture has been chosen, the training of the neural network weights does not usually guarantee that the global optimum value is achieved,

which is another drawback. This makes it difficult to estimate the robustness of the output of the neural network for small changes in the input, which is typically done by computing a Lipschitz continuity constant [5], [6].

One of the first papers to address convex neural networks was reference [7]. More recently, the training of quadratic neural networks (QNN) was shown in reference [8] to be a convex optimization program with guarantees of achieving the global optimum value of the weights. Additionally, two-layer quadratic neural networks also offer advantages relative to many other issues mentioned in the previous paragraph. In particular, the input and output are related by a quadratic form and the architecture of the network is a by-product of the training itself. This comes at the expense of only having one hidden layer, although extensions to more hidden layers have been proposed by the same authors of reference [8].

The results of reference [8] are very promising and encouraging to the community of machine learning. However, they do not address the problems of system identification and controller design with stability guarantees. The main focus of this paper is to address these two issues using quadratic neural networks. In particular, we propose convex formulations of system identification and control problems for nonlinear systems using two-layer quadratic neural networks. The paper is organized as follows. Section II will review quadratic neural networks and provide new results on their relationship with quadratic forms. System identification appears in section III and control design in section IV. Two examples are presented in section V.

II. QUADRATIC NEURAL NETWORKS

The quadratic neural network proposed in [8] is constrained to have a single hidden layer with $M = \sum_{k=1}^p M_k$ neurons. Each output is connected to M_k neurons in the hidden layer. Each output $k = 1, \dots, p$, is

$$\hat{y}^k(x) = \hat{f}^k(x) = \sum_{j=1}^{M_k} \sigma(x^T w^j) \alpha_j^k \quad (1)$$

where the activation function is quadratic and is written as

$$\sigma(z) = az^2 + bz + c \quad (2)$$

where $a \neq 0, b, c$, are pre-defined constants that parameterize the quadratic activation function. The weights w^j connect the input $x \in \mathbb{R}^n$ to each neuron j in the hidden layer whereas the weights α_j^k connect each neuron j to the output k . The

desired (label) outputs will be denoted by y and the actual outputs of the network will be denoted by \hat{y} . Following [8] it will be assumed that the weights w^j are normalized to have unit norm. Using a convex loss function $l(\cdot)$, the primal non-convex training problem for a quadratic network where all hidden neurons are connected to all outputs is [8]

$$\begin{aligned} & \min_{w^j, \alpha_j} l(\hat{y} - y) + \beta \sum_{i=1}^M \|\alpha_i\|_1 \\ & \text{s.t. } \hat{y}^k = \sum_{j=1}^M \sigma(x^T w^j) \alpha_j^k, \\ & \|w^j\|_2 = 1, \quad k = 1, \dots, p, \quad j = 1, \dots, M, \end{aligned} \quad (3)$$

for fixed $a \neq 0, b, c$, and a fixed regularization coefficient $\beta \geq 0$. The following result from reference [8] recasts the training as an equivalent convex optimization problem.

Lemma 1: [8] Given fixed $a \neq 0, b, c$, and a fixed regularization coefficient $\beta \geq 0$, the solution of the convex problem that is dual to (3) and is formulated as

$$\begin{aligned} & \min l(\hat{y} - y) + \beta \sum_{k=1}^p (Z_+^{k,4} + Z_-^{k,4}) \\ & \text{s.t. } \hat{y}_i^k = \\ & = \bar{x}_i^T \begin{bmatrix} a(Z_+^{k,1} - Z_-^{k,1}) & \frac{b}{2}(Z_+^{k,2} - Z_-^{k,2}) \\ \frac{b}{2}(Z_+^{k,2} - Z_-^{k,2})^T & c \mathbf{Trace}(Z_+^{k,1} - Z_-^{k,1}) \end{bmatrix} \bar{x}_i \\ & Z_+^{k,4} = \mathbf{Trace}(Z_+^{k,1}), \quad Z_-^{k,4} = \mathbf{Trace}(Z_-^{k,1}) \\ & Z_+^k = \begin{bmatrix} Z_+^{k,1} & Z_+^{k,2} \\ (Z_+^{k,2})^T & Z_+^{k,4} \end{bmatrix}, \quad Z_-^k = \begin{bmatrix} Z_-^{k,1} & Z_-^{k,2} \\ (Z_-^{k,2})^T & Z_-^{k,4} \end{bmatrix} \\ & Z_+^k \geq 0, \quad Z_-^k \geq 0, \quad \bar{x}_i^T = [x_i^T \quad 1] \end{aligned} \quad (4)$$

for $k = 1, \dots, p$, $i = 1, \dots, N$, where $l(\cdot)$ is a convex loss function, provides a global optimal solution for the parameters $Z_+^k, Z_-^k \in \mathbb{R}^{(n+1) \times (n+1)}$, when $M \geq M_*$ with

$$M_* = \sum_{k=1}^p [\text{rank}(Z_+^{k*}) + \text{rank}(Z_-^{k*})], \quad (5)$$

where Z_+^{k*} and Z_-^{k*} for $k = 1, \dots, p$, are the solution of the optimization problem (4) given N input data vectors $x_i \in \mathbb{R}^n$ with corresponding labels $y_i \in \mathbb{R}^p$. Moreover, the optimal value of the solutions of problems (3) and (4) are the same and therefore the duality gap is zero. \square

Instead of describing the neural network by its weights, we use the quadratic form (6), which is equivalent to the quadratic form in (4) and to the expression (1) (see [8])

$$\hat{y}^k = \hat{f}^k(x) = \bar{x}^T \begin{bmatrix} aZ_1^k & \frac{b}{2}Z_2^k \\ \frac{b}{2}(Z_2^k)^T & cZ_4^k \end{bmatrix} \bar{x} = \bar{x} \bar{Z}^k \bar{x} \quad (6)$$

where $\bar{x} = [x^T \quad 1]^T$.

Lemma 2: [9] Given a symmetric matrix P and a random vector x_* with mean μ and covariance matrix Σ ,

$$E[x_*^T P x_*] = \mu^T P \mu + \mathbf{Trace}(P \Sigma). \quad (7)$$

Proof: See Appendix B of [9]. \blacksquare

Theorem 1: Let $\bar{x} = [x^T \quad 1]^T$ with $x \in \mathbb{R}^n$, and let $f^k(z) = z^T \bar{Z}^k z$ with $\bar{Z}^k = (\bar{Z}^k)^T \in \mathbb{R}^{(n+1) \times (n+1)}$. Given parameters $a \neq 0, b, c$, the quadratic form $f^k(z)$

evaluated at $z = \bar{x}$ represents an output of a quadratic neural network with activation function (2) if and only if $\bar{Z}_{n+1, n+1}^k = \frac{c}{a} \mathbf{Trace}(\bar{Z}_{1:n, 1:n}^k)$, where $\bar{Z}_{n+1, n+1}^k$ is the $(n+1)$ -th diagonal element of \bar{Z}^k , and $\bar{Z}_{1:n, 1:n}^k \in \mathbb{R}^{n \times n}$ is the top left block submatrix of \bar{Z}^k .

Proof: The proof of the only if statement follows trivially from expression (6). To prove the if statement we assume that a quadratic form is given. Since we know a, b, c , we can then compute the values of $Z_+^{k,1} - Z_-^{k,1}$, $Z_+^{k,2} - Z_-^{k,2}$, and $Z_+^{k,4} - Z_-^{k,4}$ in expression (6). Additionally, we are assuming that $Z_+^{k,4} - Z_-^{k,4} = \mathbf{Trace}[Z_+^{k,1} - Z_-^{k,1}]$. What remains to prove is that there is no additional constraint relating $Z_+^{k,2} - Z_-^{k,2}$ with $Z_+^{k,4} - Z_-^{k,4}$ and $Z_+^{k,1} - Z_-^{k,1}$ for the quadratic form to belong to the feasible set of the optimization (4). By the Schur complement the inequality constraints from (4) are equivalent to

$$Z_+^{k,4} = \mathbf{Trace}(Z_+^{k,1}) \geq 0,$$

$$\left[1 - \mathbf{Trace}(Z_+^{k,1}) \mathbf{Trace}^\dagger(Z_+^{k,1})\right] (Z_+^{k,2})^T = 0,$$

$$Z_+^{k,1} - Z_+^{k,2} \mathbf{Trace}^\dagger(Z_+^{k,1}) (Z_+^{k,2})^T \geq 0,$$

where, defining $t_k^+ = \mathbf{Trace}(Z_+^{k,1})$,

$$\mathbf{Trace}^\dagger(Z_+^{k,1}) = \begin{cases} 0, & \text{if } t_k^+ = 0, \\ \mathbf{Trace}^{-1}(Z_+^{k,1}), & \text{if } t_k^+ \neq 0, \end{cases}$$

is the Moore-Penrose pseudo-inverse of $\mathbf{Trace}(Z_+^{k,1})$, with the same conditions applying for the case of $Z_-^{k,1}$, $Z_-^{k,2}$. Simple algebraic manipulations then lead to

$$Z_+^{k,2} (Z_+^{k,2})^T \leq Z_+^{k,1} [\mathbf{Trace}(Z_+^{k,1})],$$

$$Z_-^{k,2} (Z_-^{k,2})^T \leq Z_-^{k,1} [\mathbf{Trace}(Z_-^{k,1})].$$

Applying the trace operator to these constraints leads to

$$\begin{aligned} & \|Z_+^{k,2} - Z_-^{k,2}\|^2 \leq (\|Z_+^{k,2}\| + \|Z_-^{k,2}\|)^2 \\ & \leq [\mathbf{Trace}(Z_+^{k,1}) + \mathbf{Trace}(Z_-^{k,1})]^2 = [Z_+^{k,4} + Z_-^{k,4}]^2. \end{aligned} \quad (8)$$

We thus observe that $\|Z_+^{k,2} - Z_-^{k,2}\|$ is constrained by $Z_+^{k,4} + Z_-^{k,4}$. Note however that although the value of $Z_+^{k,4} - Z_-^{k,4}$ is constrained to be fixed given a quadratic form written as (6) and parameters a, b, c , the value of $Z_+^{k,4} + Z_-^{k,4}$ is arbitrary. Therefore, from (8) we see that there is no constraint relating $Z_+^{k,1} - Z_-^{k,1}$ or $Z_+^{k,4} - Z_-^{k,4}$ with $Z_+^{k,2} - Z_-^{k,2}$. \square

Theorem 2: Each output of a quadratic neural network satisfies the following Lipschitz inequality $\forall x_1, x_2 \in \mathbb{R}^n$

$$|\hat{f}^k(x_1) - \hat{f}^k(x_2)| \leq L_n(\bar{x}_1, \bar{x}_2, \bar{Z}^k) \|\bar{x}_1 - \bar{x}_2\|_2, \quad (9)$$

$$L_n(\bar{x}_1, \bar{x}_2, \bar{Z}^k) = \sqrt{n+1} |\lambda_{\max}(\bar{Z}^k)| (\|\bar{x}_1\|_\infty + \|\bar{x}_2\|_\infty).$$

Proof: Using the expression (6) one can write

$$|\hat{f}^k(x_1) - \hat{f}^k(x_2)| = |(\bar{x}_1 + \bar{x}_2)^T \bar{Z}^k (\bar{x}_1 - \bar{x}_2)|.$$

Using the Cauchy-Schwartz and triangular inequalities,

$$|\hat{f}^k(x_1) - \hat{f}^k(x_2)| \leq (\|\bar{x}_1\|_2 + \|\bar{x}_2\|_2) \|\bar{Z}^k\|_2 \|\bar{x}_1 - \bar{x}_2\|_2.$$

The result then follows since $\|\bar{x}\|_2 \leq \sqrt{n+1} \|\bar{x}\|_\infty$, for all $x \in \mathbb{R}^n$ and $\|\bar{Z}^k\|_2 = |\lambda_{max}(\bar{Z}^k)|$. \square

III. SYSTEM IDENTIFICATION

It is assumed that a collection of input output data pairs $\{u(k), y(k)\}_{k=1}^N$ are measured with $N \gg 1$. Based on the data one can identify the parameters of an autoregressive model of the form

$$y(t+1) = f(y(t-n+1), \dots, y(t), u(t)) \quad (10)$$

by training a quadratic neural network, where $y \in \mathbb{R}^p$, $u \in \mathbb{R}^m$, and $n \geq 1$. The value of $n-1$ gives the number of delays considered in the output. We define the training matrices as

$$X = \begin{bmatrix} u^T(n) & y^T(1) & \dots & y^T(n) \\ \vdots & \vdots & \vdots & \vdots \\ u^T(N-1) & y^T(N-n) & \dots & y^T(N-1) \end{bmatrix},$$

$$Y = [y(n+1) \dots y(N)]^T, \quad (11)$$

where $X \in \mathbb{R}^{(N-n) \times (m+pn)}$ and $Y \in \mathbb{R}^{(N-n) \times p}$. Each row of the matrix X is a neural network input sample and each row of the matrix Y is an output label of the training set of the quadratic neural network. It is assumed that $N \geq n + 0.5(pn + m + 1)(pn + m + 2)$ and that the collected data is rich enough in terms of persistent excitation [10]. After training the network, the input-output model is written as in equation (6) changing $\bar{x}(t)$ to $\bar{y}_u(t) = [u^T(t) \ y^T(t-n+1) \ \dots \ y^T(t) \ 1]^T$. Defining state variables as $x_1(t) = y(t-n+1)$, \dots , $x_n(t) = y(t)$, a state vector $x(t) = [x_1^T(t), \dots, x_n^T(t)]^T$, and noting that $\bar{y}_u(t) = [u^T(t) \ x^T(t) \ 1]^T = [u^T(t) \ \bar{x}^T(t)]^T$ yields

$$x(t+1) = Ax(t) + g(x(t), u(t)) = \begin{bmatrix} 0_{p(n-1) \times p} & I_{p(n-1) \times p(n-1)} \\ 0_{p \times p} & 0_{p \times p(n-1)} \end{bmatrix} x(t) + \begin{bmatrix} 0_{p(n-1) \times 1} \\ \mathcal{Z}(x(t), u(t)) \end{bmatrix},$$

$$\mathcal{Z}(x(t), u(t)) = \begin{bmatrix} y^1(t+1) \\ \vdots \\ y^p(t+1) \end{bmatrix} = \begin{bmatrix} \bar{y}_u^T(t) \bar{Z}^1 \\ \vdots \\ \bar{y}_u^T(t) \bar{Z}^p \end{bmatrix} \bar{y}_u(t),$$

$$\bar{Z}^i = \begin{bmatrix} \bar{Z}_{uu}^i & \bar{Z}_{ux}^i & \bar{Z}_u^i \\ (\bar{Z}_{ux}^i)^T & \bar{Z}_{xx}^i & \bar{Z}_x^i \\ (\bar{Z}_u^i)^T & (\bar{Z}_x^i)^T & \bar{Z}_{nn}^i \end{bmatrix}, \quad (12)$$

for $i = 1, \dots, p$. Expanding the quadratic forms in $\mathcal{Z}(x(t), u(t))$, the system (12) can be rewritten as

$$\bar{x}(t+1) = \bar{A}(x(t))\bar{x}(t) + \bar{B}(x(t))u(t) + E(u(t))u(t), \quad (13)$$

where $x \in \mathbb{R}^{n_x}$, $n_x = pn$, $u \in \mathbb{R}^m$, $\bar{A}(x(t)) = \mathcal{A} + F(x(t))$,

$$\mathcal{A} = \begin{bmatrix} A & 0_{pn \times 1} \\ 0_{1 \times pn} & 1 \end{bmatrix},$$

$$A = \begin{bmatrix} 0_{p(n-1) \times p} & I_{p(n-1) \times p(n-1)} \\ 0_{p \times p} & 0_{p \times p(n-1)} \end{bmatrix},$$

$$\bar{Z}_{\bar{x}\bar{x}}^i = \begin{bmatrix} \bar{Z}_{xx}^i & \bar{Z}_x^i \\ (\bar{Z}_{xx}^i)^T & \bar{Z}_{nn}^i \end{bmatrix}, \bar{Z}_{u\bar{x}}^i = [\bar{Z}_{ux}^i \ \bar{Z}_u^i],$$

$$F(x(t)) = \begin{bmatrix} 0_{p(n-1) \times (pn+1)} \\ \bar{x}^T(t) \bar{Z}_{\bar{x}\bar{x}}^1 \\ \vdots \\ \bar{x}^T(t) \bar{Z}_{\bar{x}\bar{x}}^p \\ 0_{1 \times (pn+1)} \end{bmatrix},$$

$$E(u(t)) = \begin{bmatrix} 0_{p(n-1) \times m} \\ u^T(t) \bar{Z}_{uu}^1 \\ \vdots \\ u^T(t) \bar{Z}_{uu}^p \\ 0_{1 \times m} \end{bmatrix},$$

$$\bar{B}(x(t)) = \begin{bmatrix} B(x(t)) \\ 0_{1 \times m} \end{bmatrix} = 2 \begin{bmatrix} 0_{p(n-1) \times m} \\ \bar{x}^T(t) [\bar{Z}_{u\bar{x}}^1]^T \\ \vdots \\ \bar{x}^T(t) [\bar{Z}_{u\bar{x}}^p]^T \\ 0_{1 \times m} \end{bmatrix}$$

IV. LYAPUNOV CONTROL

We assume a model of the form (13), which includes but is not limited to quadratic neural networks under the formulation (12). For a system or a neural network modelled by equation (13) we propose to design a controller

$$u(t) = K(x(t))(x(t) - x_*) + u_* = \bar{K}_{x_t} \bar{x}(t) \quad (14)$$

to stabilize the closed-loop system to a desired state x_* , where u_* is the input steady state value when $x(t) = x_*$,

$$\bar{K}_{x_t} = [K_{x_t} \ u_* - K_{x_t} x_*] \quad (15)$$

where $K_{x_t} = K(x(t))$. Replacing the control input (14) in (13) yields

$$\bar{x}(t+1) = A_{cl}(x(t), \bar{K}_{x_t})\bar{x}(t), \quad (16)$$

where

$$A_{cl}(x(t), \bar{K}_{x_t}) = \bar{A}(x(t)) + \bar{B}(x(t))\bar{K}_{x_t} + E(\bar{K}_{x_t}\bar{x}(t))\bar{K}_{x_t}. \quad (17)$$

We first design u_* based on the desired steady state response of the system and then design K_{x_t} using Lyapunov theory.

A. Steady State Input

When a system model is in the form (13) and the desired setpoint for the steady state x_* is given, one must solve

$$\bar{x}_* = \bar{A}(x_*)\bar{x}_* + \bar{B}(x_*)u_* + E(u_*)u_* \quad (18)$$

to determine the steady state value of the input u_* .

Assumption 1: The input $u_* = 0$ is a solution of (18) when $x_* = 0$.

For the case where the system model is obtained from an input-output system identification and is in the form (12) with $x(t) = [y^T(t-n+1) \ \dots \ y^T(t)]^T$, we assume that a desired setpoint y_* for the output is given. The desired

state setpoint will then be $x_* = \Gamma y_*$, where $\Gamma \in \mathbb{R}^{np \times p}$ is $\Gamma = [I_p^T \dots I_p^T]^T$, where I_p is the identity matrix of order p . The steady state value of each output is $y_{ss}^i = y_*^i$, where y_*^i is the i -th coordinate of the desired steady state output vector y_* . Therefore, from (12) one can write

$$y_*^i = \bar{y}_{u_*}^T \bar{Z}^i \bar{y}_{u_*}, \quad i = 1, \dots, p, \quad (19)$$

$$\bar{y}_{u_*}(t) = \begin{bmatrix} u_* \\ \Gamma y_* \\ 1 \end{bmatrix}. \quad (20)$$

Equations (19)–(20) must be solved for u_* given a desired setpoint y_* in order to find the steady state values of the control input. Notice from (19)–(20) that if $\bar{Z}_{nn}^i = 0$ (no constant offset terms) then $u_* = 0$ will be a solution of (19) when $y_* = 0$ and assumption 1 will be satisfied.

B. Lyapunov Controller Synthesis

After computing a solution of (19)–(20), when one exists, replacing the input (14) in equation (12) yields

$$y^i(t) = \bar{x}^T(t) \bar{Z}_{cl}^i(\bar{K}_{x_t}) \bar{x}(t), \quad i = 1, \dots, p, \quad (21)$$

where

$$\bar{Z}_{cl}^i(\bar{K}_{x_t}) = \begin{bmatrix} \bar{Z}_{x_{cl}x_{cl}}^i & \bar{Z}_{x_{cl}}^i \\ (\bar{Z}_{x_{cl}}^i)^T & \bar{Z}_{n_{cl}}^i \end{bmatrix}, \quad i = 1, \dots, p, \quad (22)$$

with

$$\begin{aligned} \bar{Z}_{x_{cl}x_{cl}}^i &= \bar{Z}_{xx}^i + K_{x_t}^T \bar{Z}_{uu}^i + [\bar{Z}_{ux}^i]^T K_{x_t} + K_{x_t}^T \bar{Z}_{uu}^i K_{x_t}, \\ \bar{Z}_{x_{cl}}^i &= \bar{Z}_x^i + (\bar{Z}_{uu}^i K_{x_t} + \bar{Z}_{ux}^i)^T (u_* - K_{x_t} x_*) + K_{x_t}^T \bar{Z}_u^i, \\ \bar{Z}_{n_{cl}}^i &= \bar{Z}_{nn}^i + 2(u_* - K_{x_t} x_*)^T Z_u^i + \delta u_*^T \bar{Z}_{uu}^i \delta u_*, \end{aligned}$$

where $\delta u_* = (u_* - K_{x_t} x_*)$. Therefore, the output is

$$y(t) = \begin{bmatrix} \bar{x}^T(t) \bar{Z}_{cl}^1(\bar{K}_{x_t}) \\ \vdots \\ \bar{x}^T(t) \bar{Z}_{cl}^p(\bar{K}_{x_t}) \end{bmatrix} \bar{x}(t). \quad (23)$$

From (12) and (23) the closed-loop state space model can be rewritten as in equation (16) where

$$A_{cl}(x(t), \bar{K}_{x_t}) = \begin{bmatrix} \bar{I} \\ \bar{x}^T(t) \bar{Z}_{cl}^1(\bar{K}_{x_t}) \\ \vdots \\ \bar{x}^T(t) \bar{Z}_{cl}^p(\bar{K}_{x_t}) \\ \bar{0}^T \end{bmatrix}, \quad (24)$$

$\bar{I} = [0_{p(n-1) \times p} \quad I_{p(n-1) \times p(n-1)} \quad 0_{p(n-1) \times 1}]$, and $\bar{0}^T = [0_{1 \times pn} \quad 1]$. Equation (24) can be rewritten in the form (17). The next Theorem provides a Lyapunov-based design strategy that yields a provably stabilizing controller.

Theorem 3: Given a desired setpoint $x_* \in \mathcal{X} \subseteq \mathbb{R}^{n_x}$, if

$$\begin{aligned} A_{cl}^T(x(t), \bar{K}_{x_t}) \bar{P} A_{cl}(x(t), \bar{K}_{x_t}) - \bar{P} &\leq 0, \\ \forall x(t) \neq x_*, x(t) \in \mathcal{X} \subseteq \mathbb{R}^{n_x}, \end{aligned} \quad (25)$$

is satisfied where A_{cl} is given by (17) [or by (24)] and \bar{P} is defined as

$$\bar{P} = \begin{bmatrix} P & -P x_* \\ -x_*^T P & x_*^T P x_* \end{bmatrix} \quad (26)$$

where $P > 0$, then the controller $u(t) = \bar{K}_{x_t} \bar{x}(t)$ renders the closed-loop system (16) stable in the sense of Lyapunov inside the largest invariant set of the Lyapunov function (27) fully contained in \mathcal{X} . If the inequality (25) is strict, the closed-loop system is asymptotically stable inside the same invariant set. If $\mathcal{X} = \mathbb{R}^{n_x}$, the stability is global.

Proof: If x_* is the desired setpoint in steady state we define the candidate quadratic control Lyapunov function

$$V(x(t)) = (x(t) - x_*)^T P (x(t) - x_*) = \bar{x}^T(t) \bar{P} \bar{x}(t), \quad (27)$$

where \bar{P} is defined in (26). For Lyapunov stability,

$$V(x_*) = 0, \quad (28)$$

$$V(x) > 0, \quad \forall x \neq x_*, x \in \mathcal{X} \quad (29)$$

$$\Delta V \leq 0, \quad \forall x(t) \neq x_*, x(t) \in \mathcal{X}, \quad (30)$$

where $\Delta V = V(x(t+1)) - V(x(t))$. Notice from the definition of the Lyapunov function (27) that the condition (28) is guaranteed to be satisfied. Furthermore, the condition (29) is also satisfied because $P > 0$. Computing $V(x(t+1))$ using (16) yields

$$V(x(t+1)) = \bar{x}^T(t) A_{cl}^T(x(t), \bar{K}_{x_t}) \bar{P} A_{cl}(x(t), \bar{K}_{x_t}) \bar{x}(t). \quad (31)$$

Using (31) condition (30) is implied by (25). From standard Lyapunov theory the system is then stable inside the largest invariant set of the Lyapunov function (27) fully contained in \mathcal{X} . The proof of asymptotical stability follows by noting that the Lyapunov function is decreasing when the strict inequality is satisfied. Global stability follows from the fact that the Lyapunov function (27) is radially unbounded. \square

Note that, according to Theorem 1, for a known vector x_* the Lyapunov function (27) is a quadratic neural network with activation parameters $a \neq 0, b, c$, if and only if $x_*^T P x_* = \frac{c}{a} \mathbf{Trace}(P)$.

Theorem 4: Consider the quadratic Lyapunov function (27) with $P > 0$. Given scalars c and $a \neq 0$, if the desired steady state x_* is a random vector with zero mean and covariance matrix $\Sigma = \frac{c}{a} I$, where I is the identity, then

$$E[x_*^T P x_*] = \frac{c}{a} \mathbf{Trace}(P). \quad (32)$$

In other words, the matrix \bar{P} in (26) can represent "on average" a quadratic neural network if and only if x_* is a zero mean random vector with covariance matrix $\Sigma = \frac{c}{a} I$.

Proof: Use Theorem 1 and Lemma 2, $\mu = 0, \Sigma = \frac{c}{a} I$. \square

Unfortunately, condition (25) is not convex. The following assumption enables control design as a convex optimization.

Assumption 2: $E(u(t)) = 0$ for all $u(t) \in \mathbb{R}^m$.

Under assumptions 1 and 2 the model (13) becomes

$$x(t+1) = A(x(t))x(t) + B(x(t))u(t), \quad (33)$$

for appropriate matrices $A \in \mathbb{R}^{n_x \times n_x}, B \in \mathbb{R}^{n_x \times m}$. The following result can then be stated and proved.

Theorem 5: Assume that a system model is given in the form (33) where $x(t) \in \mathcal{X} \subseteq \mathbb{R}^{n_x}$. If the desired steady

state setpoint is $x_* = 0$ and if for a given $\epsilon \in (0, 1)$ there are $P = P^T$ and $L(x(t))$ satisfying

$$\begin{bmatrix} (1 - \epsilon)P - \epsilon I & (*)^T \\ A(x(t))P + B(x(t))L(x(t)) & P \end{bmatrix} \geq 0 \quad (34)$$

$\forall x(t) \in \mathcal{X}$, where $(*) = A(x(t))P + B(x(t))L(x(t))$, then the closed-loop system is asymptotically stable in the largest invariant set of the Lyapunov function $V(x) = x^T P^{-1}x$ fully contained inside \mathcal{X} . If $\mathcal{X} = \mathbb{R}^{n_x}$ then the stability is global. The control input is $u(t) = L(x(t))P^{-1}x(t)$.

Proof: Replacing $u(t) = K(x(t))x(t)$ in (33) yields

$$\begin{aligned} x(t+1) &= A_{cl}(x(t), K(x(t))x(t)), \\ A_{cl}(x(t), K(x(t))) &= A(x(t)) + B(x(t))K(x(t)). \end{aligned} \quad (35)$$

Consider a candidate Lyapunov function of the form $V(x) = x^T P^{-1}x$. Note that if the condition (34) is satisfied then $P > 0$ and therefore it is invertible. It is clear that $V(0) = 0$ and $V(x) > 0$ for $x \neq 0$. Additionally, if (34) is satisfied then applying the Schur complement and using (35) it implies that $PA_{cl}^T(x(t), K(x(t)))P^{-1}A_{cl}(x(t), K(x(t)))P - (1 - \epsilon)P \leq -\epsilon I < 0 \forall x \in \mathcal{X}$, using the substitution $K = L(x(t))P^{-1}$. Multiplying on the left by $x^T(t)P^{-1}$ and on the right by $P^{-1}x(t)$ yields $\Delta V < -\epsilon x^T P^{-1}x$, $\forall x \in \mathcal{X}$, which guarantees asymptotic stability in the largest level set of $V(x)$ fully contained in \mathcal{X} . Global stability follows because the Lyapunov function is radially unbounded. \square

C. Convex Formulation of Lyapunov Controller Synthesis

To formulate controller synthesis as a convex optimization problem we will need the framework of sum of squares polynomials. For $\mathbf{x} \in \mathbb{R}^n$, a multivariate polynomial $p(\mathbf{x})$ is a sum of squares (SOS) if there exist some polynomials $f_i(\mathbf{x})$, $i = 1, \dots, M$, such that $p(\mathbf{x}) = \sum_{i=1}^M f_i^2(\mathbf{x})$. A polynomial $p(\mathbf{x})$ of degree $2d$ is a sum of squares if and only if there exists a positive semidefinite matrix Q and a vector $W(\mathbf{x})$ containing monomials in \mathbf{x} of degree less than d such that $p(\mathbf{x}) = W(\mathbf{x})^T Q W(\mathbf{x})$ [11]. It should be noted that if $p(\mathbf{x})$ is a sum of squares then $p(\mathbf{x}) \geq 0$, but the converse is generally not true. For a convex formulation the inequality (34) will be relaxed into a sum of squares and $K(x(t))$ will be constrained to have polynomial entries.

Corollary 1: Assume that a system model is given in the form (33) where $A(x(t))$ and $B(x(t))$ have polynomial entries and $x(t) \in \mathcal{X} \subseteq \mathbb{R}^{n_x}$. If the desired steady state setpoint is $x_* = 0$ and if for a given $0 \leq \epsilon < 1$ there are $P > 0$ and $L(x(t))$ with polynomial entries satisfying

$$\begin{bmatrix} (1 - \epsilon)P - \epsilon I & (*)^T \\ A^T(x(t))P + B(x(t))L(x(t)) & P \end{bmatrix} \text{ is SOS,} \quad (36)$$

$\forall x(t) \in \mathcal{X}$, where $(*) = PA(x(t)) + L^T(x(t))B^T(x(t))$, then the closed-loop system is stable in the largest invariant set of the Lyapunov function $V(x) = x^T P^{-1}x$ fully contained inside \mathcal{X} . If the condition (36) is satisfied for $\epsilon \in (0, 1)$, then the closed-loop system is asymptotically stable inside the same invariant set. If $\mathcal{X} = \mathbb{R}^n$ the stability is global. The control input is given by $u(t) = L(x(t))P^{-1}x(t)$.

Proof: The proof follows the argument of the proof of Theorem 5 upon observing that satisfaction of the condition (36) implies that the inequality (34) is satisfied. \square

To measure performance one can consider the cost

$$J(x, u) = \sum_{k=0}^{\infty} (x^T(k)Qx(k) + u^T(k)Ru(k)) \quad (37)$$

for given matrices $Q = Q^T \geq 0$ and $R = R^T > 0$. The lower this cost is for a given controller the better is its performance.

Theorem 6: Assume that a system model is given in the form (33) where $A(x(t))$ and $B(x(t))$ have polynomial entries. If there are $P = P^T$ and $W(x(t))$ defined as (38)

$$\begin{bmatrix} A^T(x)PA(x) - P + Q & A^T(x)PB(x) \\ B^T(x)PA(x) & R + B^T(x)PB(x) \end{bmatrix} \quad (38)$$

and (38) is SOS $\forall x(t) \in \mathbb{R}^{n_x}$, then any controller that asymptotically stabilizes the system to the origin yields a cost (37) that is bounded below by $V(x_0) = x_0^T P x_0$ when the system trajectories start from an initial condition $x_0 \in \mathbb{R}^{n_x}$.

Proof: The condition (38) implies that the matrix $W(x(t))$ defined by (38) is positive semidefinite for all $x(t) \in \mathbb{R}^{n_x}$. Left multiplying $W(x(t))$ by $[x^T(t) \ u^T(t)]$ and right multiplying by $[x^T(t) \ u^T(t)]^T$ and using (33),

$$x^T(t)Qx(t) + u^T(t)Ru(t) + V(x(t+1)) - V(x(t)) \geq 0, \quad (39)$$

where $V(x(t)) = x^T(t)Px(t)$. Summing the terms on the left hand side of inequality (39) from zero to infinity yields

$$J(x, u) \geq V(x(0)) - V\left(\lim_{t \rightarrow \infty} x(t)\right) = V(x_0), \quad (40)$$

provided $x(t) \rightarrow 0$ as $t \rightarrow \infty$, which is guaranteed when the closed-loop system is asymptotically stable to the origin. \square

Remark 1: For some systems in the form (33) the optimal cost to go to the origin from a given state x is $V(x) = x^T P x$, where P is the matrix of maximum trace that satisfies condition (38). That is the case for example when the system is linear and the inequality (39) becomes an equality called the Bellman equation [12]. The optimal control is

$$u(t) = -(R + B^T(x)PB(x))^{-1} B^T(x)PA(x)x \quad (41)$$

A common heuristic is to find P with maximum trace that satisfies condition (38) and to compute the control (41).

V. EXAMPLES

Example 1: We perform system identification of a flexible robot arm using $n = 1$. The data is available at the website of the Database for the Identification of Systems (DaISy) [13]. The input was a periodic sine sweep. An output of $N = 1024$ data points was collected. The training was done for a quadratic activation function with parameters $a = 0.0937, b = 0.5, c = 0.4688$, using a regularization coefficient of $\beta = 0.01$ and an infinity norm loss function $l(\cdot)$. Only the first 122 data points (12% of all data) were used to train the neural network. The training was solved with cvx [14] using MOSEK [15] and yielded an optimal

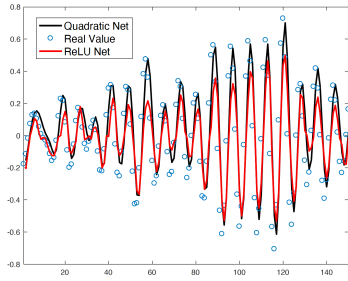


Fig. 1. QNN (black), by ReLU (red) and data points (blue circles).

value of the objective function equal to 0.49 and 6 neurons. The resulting model is $y(t+1) = \bar{y}_u^T(t)\bar{Z}\bar{y}_u(t)$ where $\bar{y}_u(t) = [u^T(t) \ y^T(t) \ 1]^T$,

$$\bar{Z} = \begin{bmatrix} -0.0989 & 0.5030 & -0.1682 \\ 0.5030 & 0.1599 & 1.8265 \\ -0.1682 & 1.8265 & 0.0610 \end{bmatrix}.$$

For comparison purposes a two layer ReLU neural network was trained with $\beta = 0.01$ yielding a similar value of the objective function, namely 0.47, but at the cost of needing 396 neurons. Figure 1 compares the predictions with the data.

Example 2: Consider a model of a quadrotor flying at a constant altitude in the positive X direction.

$$\begin{bmatrix} \dot{X} \\ \dot{V}_X \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} X \\ V_X \end{bmatrix} + \begin{bmatrix} 0 \\ g \end{bmatrix} u - \begin{bmatrix} 0 \\ d \end{bmatrix} V_X^2, \quad (42)$$

where X and V_X are the position and the velocity of the quadrotor, $u = \tan(\theta)$ is the control input where θ is the pitch angle, $g = 9.8ms^{-2}$ is the acceleration of gravity, and d is a drag-related coefficient. The control objective is to take the system to the origin from any initial position. The model (42) is discretized using a forward Euler approximation as

$$A(x(t)) = \begin{bmatrix} 1 & T \\ 0 & 1 - TdV_X(t) \end{bmatrix}, \quad B(x(t)) = \begin{bmatrix} 0 \\ Tg \end{bmatrix}.$$

For $T = 0.102s$ we get $Tg = 1ms^{-1}$ and assume $Td = 0.0023sm^{-1}$. The P with minimum condition number that satisfies (34) with $\epsilon = 0.1$ and a first order polynomial $L(x)$ is

$$P = \begin{bmatrix} 1.4589 & -1.6008 \\ -1.6008 & 2.6636 \end{bmatrix},$$

which represents a quadratic neural network with $ca^{-1} = (2.6636)(1.4589)^{-1} = 1.8258$. The controller is given by $u(t) = -1.1556X(t) - 1.1771V_X(t) + 0.0023V_X^2(t)$ and the system response for an initial condition of $X(0) = 1, V_X(0) = 0$ is shown in figure 2. The controller cancels the term in $V_X^2(t)$ in the open-loop system and yields a linear closed-loop system with stable eigenvalues $\lambda_1 = 0.8895, \lambda_2 = -0.0666$. The maximum trace P that satisfies (38) for $Q = I, R = 2.2 \times 10^{-16}$ using SoStTools [11] is

$$P = \begin{bmatrix} 11.3167 & 1.0523 \\ 1.0523 & 1.1073 \end{bmatrix},$$

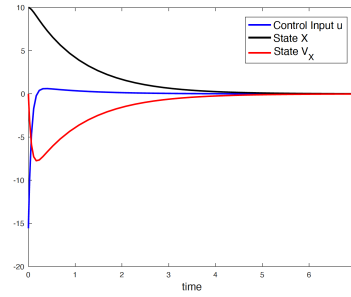


Fig. 2. State X (black), V_X (red), and control input (blue)).

which represents a quadratic neural network with $ca^{-1} = 0.0979$. Using equation (41) leads to the controller $u(t) = -0.9503X(t) - 1.097V_X(t) + 0.0023V_X^2(t)$. Note that this controller also cancels the term in $V_X^2(t)$ from the open-loop system and yields a linear closed-loop system with stable eigenvalues $\lambda_1 = 0.9031, \lambda_2 = -0.0001$. One can write

$$u(t) = \begin{bmatrix} X \\ V_X \\ 1 \end{bmatrix}^T \begin{bmatrix} 0 & 0 & -0.47515 \\ 0 & 0.0023 & -0.5485 \\ -0.47515 & -0.5485 & 0 \end{bmatrix} \begin{bmatrix} X \\ V_X \\ 1 \end{bmatrix}.$$

Thus the control can be implemented by a QNN with $c = 0$.

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